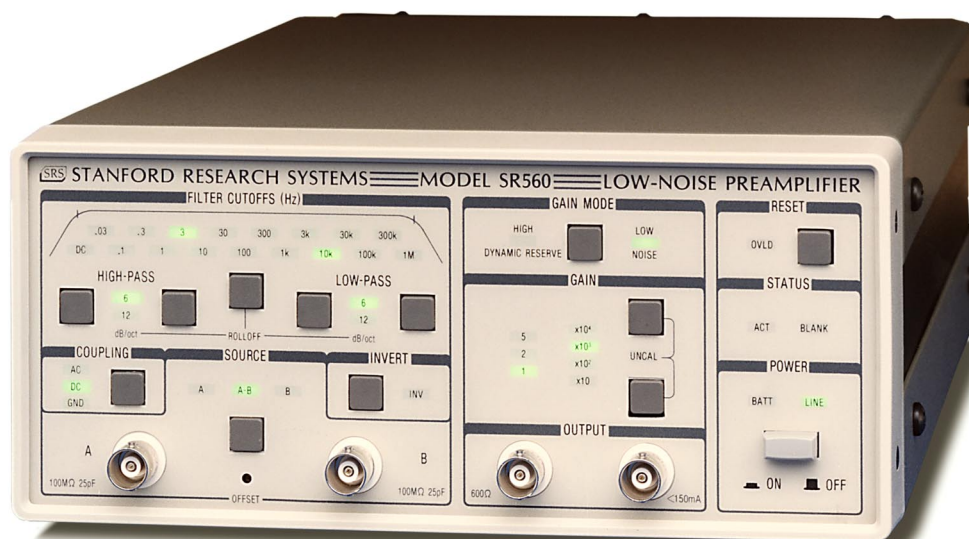


# Low-Noise Voltage Preamplifier

SR560 — DC to 1 MHz voltage preamplifier



## SR560 Low-Noise Preamplifier

- **4 nV/√Hz input noise**
- **1 MHz bandwidth**
- **Variable gain from 1 to 50,000**
- **AC or DC coupled**
- **Two configurable signal filters**
- **Differential and single-ended inputs**
- **Line or battery operation**
- **RS-232 interface**

The SR560 is a high-performance, low-noise, general purpose preamplifier. With an input noise of only 4 nV/√Hz, a 1 MHz bandwidth and a gain of up to 50,000, the SR560 is ideal for a wide variety of applications including low temperature measurements, optical detection and audio engineering.

### Inputs

The SR560 has a differential front end with 4 nV/√Hz input noise and an input impedance of 100 MΩ. Complete noise figure contours are shown in the SR560 Tech Note at the end of this section. The SR560's inputs are fully floating (BNC shields are not connected to chassis ground). Both the amplifier ground and the chassis ground are available on the rear panel for flexibility in grounding the instrument. Input offset nulling is accomplished by a front-panel potentiometer, accessible with a small screwdriver.

In addition to the signal inputs, a rear-panel TTL blanking input lets you quickly turn off and on the instrument's gain. This is useful in preventing front-end overloading. The gain turns off 5 μs after the TTL level goes high, and back on again within 10 μs after the TTL signal goes low.

### Outputs

Two insulated output BNC connectors provide 600 Ω and 50 Ω outputs. Both are capable of driving 10 Vpp into their respective loads. Two rear-panel power supply outputs

provide up to 200 mA of  $\pm 12$  VDC referenced to the amplifier ground. The outputs provide clean DC power for use as a bias source.

### Gain

Gain is selectable from 1 to 50,000 in a 1–2–5 sequence. An adjustable gain feature lets you specify the gain as a percentage of any of the fixed gain settings with 0.5 % resolution. Gain can be selectively allocated before the filters to optimize noise performance, or after the filters to reduce susceptibility to overloads.

### Filters

The SR560 contains two, first-order RC filters whose cutoff frequency and type (HPF or LPF) can be configured from the front panel. Together, the filters can be configured as a 6 or 12 dB/oct rolloff low pass or high pass filter, or as a 6 dB/oct rolloff band pass filter. A filter reset button is included to shorten the overload recovery time of the instrument when long filter time constants are being used. Filter cutoff frequencies can be set in a 1–3–10 sequence from 0.03 Hz to 1 MHz.

### Battery Operated

Three internal, rechargeable, lead-acid batteries provide up to 15 hours of battery powered operation. An internal battery charger automatically charges the batteries when the unit is connected to the line. The charger senses the battery state and adjusts the charging rate accordingly. Two rear-panel LEDs indicate the charge state of the batteries. When the batteries become discharged, they are automatically disconnected from the amplifier circuit to avoid battery damage.

### No Digital Noise

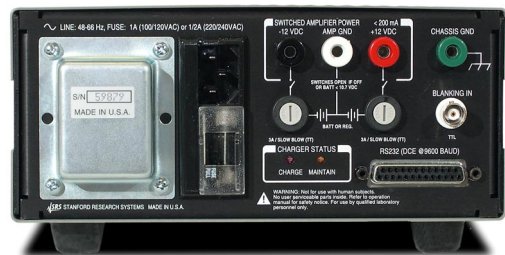
The microprocessor that runs the SR560 is "asleep" except during the brief interval it takes to change the instrument's setup. This ensures that no digital noise will contaminate your low-level analog signals.

### RS-232 Interface

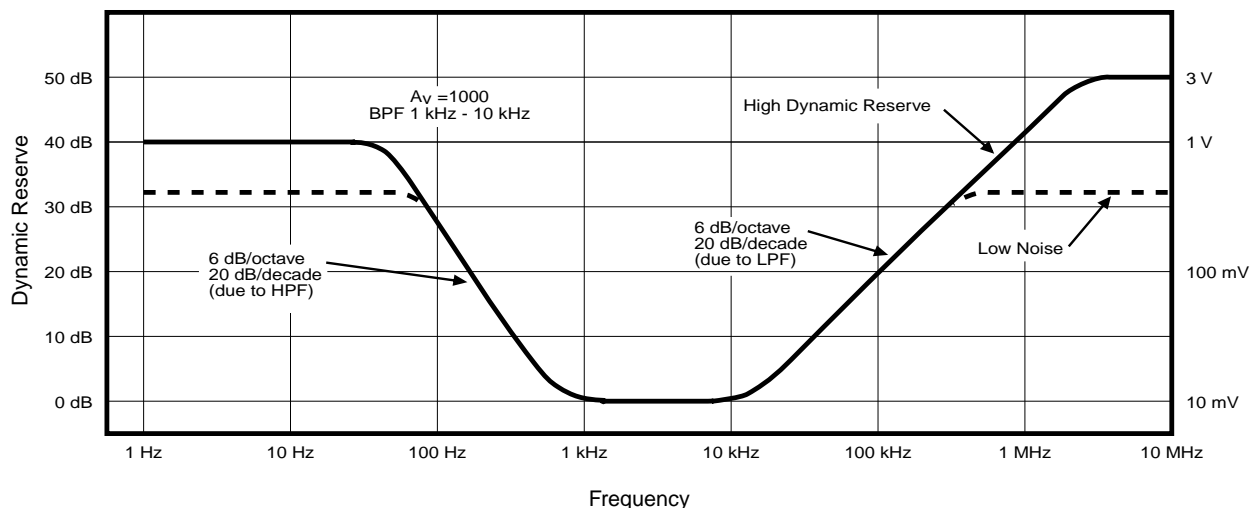
The RS-232 interface allows listen-only communication with the SR560 at 9600 baud. Up to four SR560s can be controlled from a single computer, with each SR560 being assigned a unique address. A "Listen" command specifies which SR560 will respond to commands on the RS-232 line. All functions of the instrument (except power on) can be set via the RS-232 interface. The RS-232 interface electronics are opto-isolated from the amplifier circuitry to provide maximum noise immunity.

### Ordering Information

SR560	Low-noise voltage preamplifier
O560RMD	Double rack mount kit
O560RMS	Single rack mount kit
O560SB	Spare battery set (3 batteries)



SR560 rear panel



**Input**

Inputs	AC or DC coupled, single-ended or differential
Input impedance	100 M $\Omega$ + 25 pF
Maximum input	3 V <sub>pp</sub>
CMRR	100 dB from DC to 1 kHz. Decreases by 6 dB/octave from 1 kHz to 1 MHz.
Noise	4 nV/ $\sqrt{\text{Hz}}$ at 1 kHz
Gain	1 to 50,000 in 1–2–5 sequence Vernier gain in 0.5 % steps
Gain stability	200 ppm/ $^{\circ}\text{C}$
Frequency response	$\pm 0.5$ dB to 1 MHz, $\pm 0.3$ dB to 300 kHz (gains up to 1000)

**Filters**

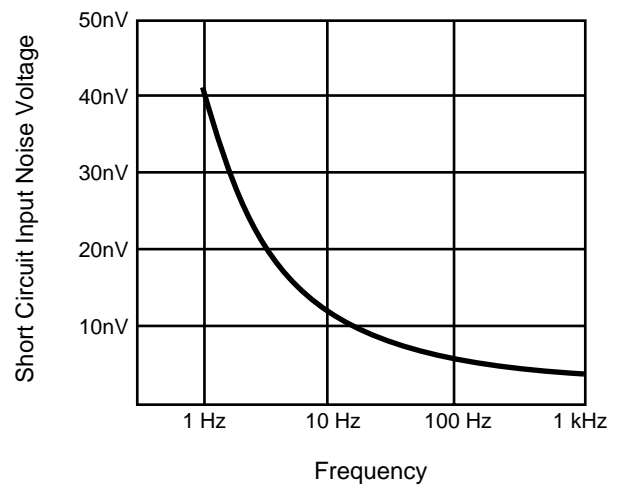
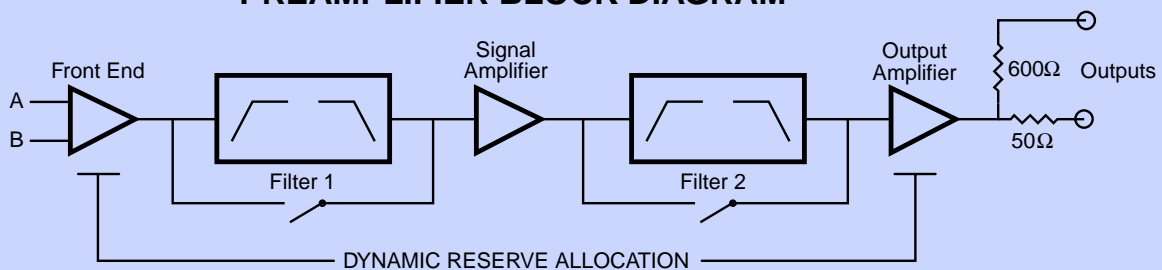
Signal filters	2 configurable (low-pass or high-pass) 6 dB/oct rolloff filters. –3 dB points are settable in a 1–3–10 sequence from 0.03 Hz to 1 MHz.
Gain allocation	High Dynamic Reserve—Gain is increased after the signal filters to prevent overloading. Low Noise—Gain is increased before the filters to improve noise figure.

**Output**

Maximum output	10 V <sub>pp</sub> into 50 $\Omega$ and 600 $\Omega$
Filter reset	Long time constant filters may be reset with front-panel button.
DC drift	5 $\mu\text{V}/^{\circ}\text{C}$ referred to input (DC coupled)
Distortion	0.01 % at 1 kHz
Rear panel	$\pm 12$ VDC @ 200 mA referenced to amplifier ground

**General**

External gating	TTL input sets gain to zero
Interfaces	RS-232, 9600 baud, receive only
Power	100/120/220/240 VAC. 6 watts charged, 30 watts while charging. Internal batteries provide 15 hours of operation between charges. Batteries are charged while connected to the line.
Dimensions	8.3" $\times$ 3.5" $\times$ 13.0" (WHD)
Weight	15 lbs. (batteries installed)
Warranty	One year parts and labor on defects in materials and workmanship


**PREAMPLIFIER BLOCK DIAGRAM**


## Noise Figure Contours

Noise figure contours are often provided with amplifier specifications but many users are unclear on what they signify. The noise figure of an amplifier is the ratio (usually expressed in dB) of the equivalent input noise of the amplifier at a given frequency to the thermal noise of a source with a given source impedance. Since the equivalent input noise of an amplifier is simply the output noise divided by the gain, this can be expressed as:

$$NF = 20 \log (\text{Output Noise/Gain})/(\text{Source Thermal Noise})$$

If the amplifier were noiseless, all the output noise would be due to the source thermal noise, and the noise figure would be 0 dB. To the extent that the amplifier adds some of its own noise, the Output Noise/Gain will be bigger than the source thermal noise and the noise figure will be non-zero. Thus, the noise figure is an indication of how much of the output noise is contributed by the amplifier.

Since amplifier noise and thermal noise are functions of both frequency and source impedance, noise figures are often

plotted as contours on a graph of source resistance versus frequency. At a given frequency, the noise figure is large for very low source resistance since the thermal noise of the source is small and any amplifier noise leads to a large noise figure. As the source resistance increases, the noise figure becomes better until the current noise of the amplifier passing through the high source resistance once again degrades the noise figure.

At a constant source resistance, the noise figure is poor at low frequencies because of the large  $1/f$  noise of the amplifier. As the frequency is increased,  $1/f$  noise decreases and the noise figure improves. Eventually, for all transistors, the current noise increases as a function of frequency, and the noise figure begins to increase again.

Does this mean that you should always increase the resistance of your source to reach the region of minimum noise figure? Certainly not. By doing that you would only be making your source noisier to improve the noise figure. Remember, the goal is to reduce total noise, not to minimize the noise figure.

